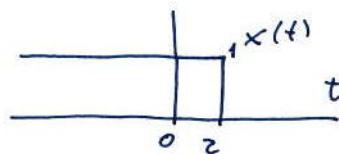
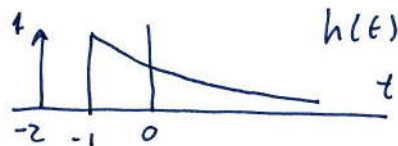


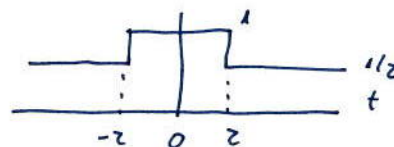
$$x(t) = u(2-t)$$



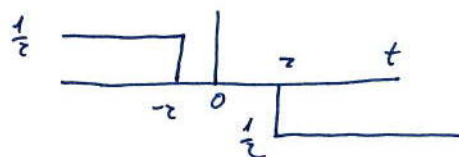
$$h(t) = e^{-\tau t} u(t+1) + \delta(t+2)$$



$$a) E_v \{x(t)\} = \frac{x(t) + x(-t)}{2}$$



$$O_d \{x(t)\} = \frac{x(t) - x(-t)}{2}$$



$$P_i(t) = u(2-t)$$

$$E = \int_{-\infty}^{\infty} u(2-t) dt = \int_{-\infty}^2 1 \cdot dt = \infty$$

$$P_{AV} = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T dt = \lim_{T \rightarrow \infty} \frac{2+T}{2T} = \frac{1}{2}$$

$$b) h(t) \neq k \delta(t) \Rightarrow \text{CON MEMORIA}$$

$$h(t) \neq 0 \begin{cases} t > 0 \\ t < 0 \end{cases} \Rightarrow \text{ANTICAUSAL}$$

$$\int_{-\infty}^{\infty} |h(t)| dt = \int_{-\infty}^{\infty} \delta(t+2) dt + \int_1^{\infty} e^{-\tau t} dt = 1 + \frac{e^{-\tau}}{\tau} < \infty \Rightarrow \text{ESTABILE}$$

$$c) y(t) = x(t) * h(t) \Rightarrow h(t) = \delta(t+2) + \underbrace{e^{-\tau t} u(t+1)}_{h_1(t)}$$

$$= x(t) * (\delta(t+2) + h_1(t))$$

$$= x(t+2) + \underbrace{x(t) * h_1(t)}_{y_1(t)} \rightarrow x(t+2) = u(-t)$$

$$y_1(t) = \int_{-\infty}^{\infty} e^{-\tau(t-\tau)} u(t+1-\tau) d\tau$$

$$y_1(t) = \begin{cases} t < 1 & \int_{-\infty}^{t+1} e^{-zt} e^{z\tau} d\tau = \frac{e^z}{z} \\ t \geq 1 & \int_{-\infty}^z e^{-zt} e^{z\tau} d\tau = \frac{e^{-z(t-z)}}{z} \end{cases}$$

$$y(t) = \begin{cases} 1 + \frac{e^z}{z} & t < 0 \\ \frac{e^z}{z} & 0 \leq t < 1 \\ \frac{e^{-z(t-1)}}{z} & t \geq 1 \end{cases}$$

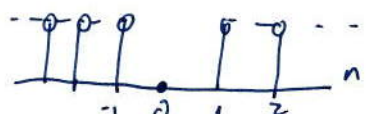
$$y(t) = u(-t) + \frac{e^z}{z} u(1-t) + \frac{e^{-z(t-1)}}{z} u(t-1)$$

d) i) PERIÓDICA $T = \frac{2\pi}{\sqrt{z}}$

ii) ~~e^{-znt}~~ $e^{-znt} \Rightarrow$ es una exp. real, por lo tanto
NO PERIÓDICA

iii) $\omega = \frac{1}{z}$ $\frac{\omega_0}{2\pi} \notin \mathbb{Q}$ NO PERIÓDICA

iv) $\omega_0 = 1$ $\frac{\omega_0}{2\pi} \notin \mathbb{Q}$ NO PERIÓDICA

v)  $\Rightarrow$ NO PERIÓDICA

a) $x(t) = \sin(\pi t)$

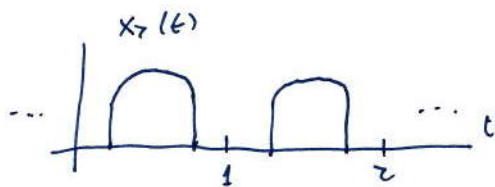
→ La señal ya es una serie de Fourier, ya que se puede escribir como suma de exponenciales complejas

$$x(t) = \frac{1}{2j} e^{j\pi t} - \frac{1}{2j} e^{-j\pi t} = \underbrace{\frac{1}{2j} e^{j\frac{\pi}{2}t}}_{k=1} - \underbrace{\frac{1}{2j} e^{-j\frac{\pi}{2}t}}_{k=-1}$$

De aquí $C_k = \begin{cases} \frac{1}{2j} & k=1 \\ -\frac{1}{2j} & k=-1 \\ 0 & k \neq \pm 1 \end{cases} \quad T=2$

Serie de Fourier: $x(t) = \sum_k C_k e^{jk\pi t}$

b) La señal $x_2(t)$ es periódica de periodo $T=1$



La S. Fourier: $x(t) = \sum_{k=-\infty}^{\infty} C_k e^{jk2\pi t}$

$$C_k = \int_{1/6}^{5/6} \sin(\pi t) e^{-jk2\pi t} dt = \int_{1/6}^{5/6} \frac{1}{2j} (e^{j\pi t} - e^{-j\pi t}) e^{-jk2\pi t} dt$$

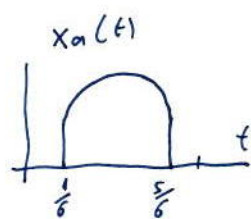
$$= \frac{1}{2\pi} \left[\frac{e^{j\frac{\pi}{6}(1-2k)}}{(1-2k)} + \frac{e^{-j\frac{\pi}{6}(1+2k)}}{(1+2k)} - \frac{e^{j\frac{5\pi}{6}(1-2k)}}{(1-2k)} - \frac{e^{-j\frac{5\pi}{6}(1+2k)}}{(1+2k)} \right]$$

$\forall k \in \mathbb{Z}$

Podrían simplificarse los coeficientes: $(e^{-j\frac{\pi}{3}k} = e^{-j(\frac{\pi}{3}+2\pi)k} = e^{j\frac{5\pi}{3}k})$

$$C_k = \frac{1}{\pi(1-4k^2)} \left[\sqrt{3} \cos\left(\frac{5\pi}{3}k\right) - 2k \sin\left(\frac{5\pi}{3}k\right) \right]$$

De manera alternativa, se puede calcular a partir de la T.F. de la señal aperiódica:



$$x_a(t) = \text{sen}(\pi t) \cdot x_1(t)$$

$$\text{con } x_1(t) = \begin{cases} 1 & \frac{1}{6} < t < \frac{5}{6} \\ 0 & \text{resto} \end{cases}$$

$$X_a(\omega) = \frac{1}{2\pi} X(\omega) * \sum_i \pi [\delta(\omega - \pi) - \delta(\omega + \pi)]$$

$$X_1(\omega) = \frac{2 \text{sen}(\frac{\omega}{3})}{\omega} \cdot e^{-j\frac{\omega}{3}}$$

$$X_a(\omega) = e^{-j\frac{\omega}{2}} \left[\frac{\text{sen}(\frac{\omega - \pi}{3})}{\omega - \pi} - \frac{\text{sen}(\frac{\omega + \pi}{3})}{\omega + \pi} \right]$$

$$C_k = X_a(2\pi k) = \frac{(-1)^k}{\pi} \left[\frac{\text{sen}(\frac{2\pi}{3}k - \frac{\pi}{3})}{2k - 1} - \frac{\text{sen}(\frac{2\pi}{3}k + \frac{\pi}{3})}{2k + 1} \right]$$

$$k \in \mathbb{Z}$$

(Aunque parezca distinto a la solución anterior, se puede demostrar que es la misma)

c) Sabemos que $e^{j\omega_0 t} \rightarrow \boxed{h(t)} \rightarrow H(\omega_0) e^{j\omega_0 t}$

En el caso de una señal periódica, la SF a la entrada da una SF a la salida:

$$x(t) = \sum_k C_k e^{jk \frac{2\pi}{T} t} \rightarrow \boxed{h(t)} \rightarrow y(t) = \sum_k \underbrace{C_k H(k \frac{2\pi}{T})}_{d_k} e^{jk \frac{2\pi}{T} t}$$

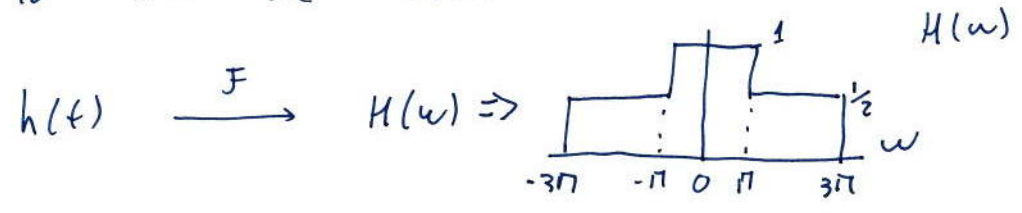
En nuestro caso:

$$y(t) = \sum_k \underbrace{C_k \cdot H(k 2\pi)}_{d_k} e^{jk \frac{2\pi}{T} t}$$

La S.F. a la salida tendrá el mismo periodo $T=1$
y los coef. de serán

$$d_k = C_k H(k\pi)$$

Calcule la T.F. de $h(t)$



Por lo tanto $H(k\pi) = \begin{cases} 1 & k=0 \\ \frac{1}{2} & k=\pm 1 \\ 0 & |k| > 1 \end{cases}$

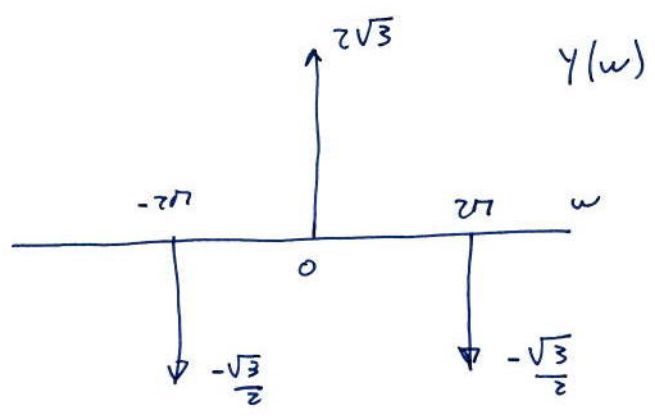
De aquí:

$$d_k = \begin{cases} C_0 & k=0 \\ \frac{C_1}{2} & k=1 \\ \frac{C_{-1}}{2} & k=-1 \\ 0 & |k| > 1 \end{cases}$$

$$y(t) = C_0 + \frac{C_1}{2} e^{j\pi t} + \frac{C_{-1}}{2} e^{-j\pi t}$$

d) La T.F. de una señal periódica es de la forma

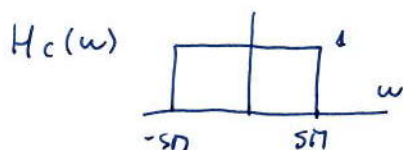
$$Y(\omega) = \sum_k z_n d_k \delta(\omega - 2\pi k) \Rightarrow \begin{cases} C_0 = \frac{\sqrt{3}}{\pi} \\ C_1 = C_{-1} = -\frac{\sqrt{3}}{2\pi} \end{cases}$$



TAMBIÉN PUEDE HACERSE COMO $Y(\omega) = X(\omega) \cdot H(\omega)$

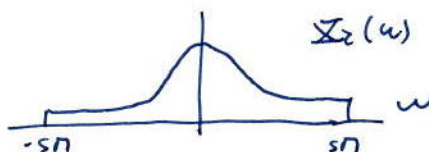
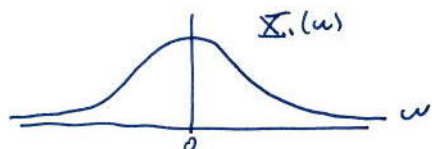
3

$$a) \quad x_1(t) = e^{-2|t|} \xrightarrow{F} \quad X_1(\omega) = \frac{1}{2+j\omega} + \frac{1}{2-j\omega} = \frac{4}{4+\omega^2}$$



$$X_2(\omega) = X_1(\omega) \cdot H_c(\omega)$$

$$= \begin{cases} \frac{4}{4+\omega^2} & |\omega| \leq 5\pi \\ 0 & |\omega| > 5\pi \end{cases}$$



$x_1(t)$ no es muestreable, ya que $X_1(\omega)$ no es de banda limitada.

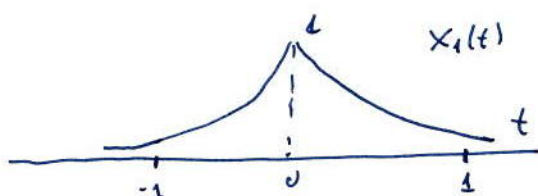
$x_2(t)$ sí es muestreable. $\omega_{\max} = 5\pi$

Podemos muestrear a la freq. de Nyquist

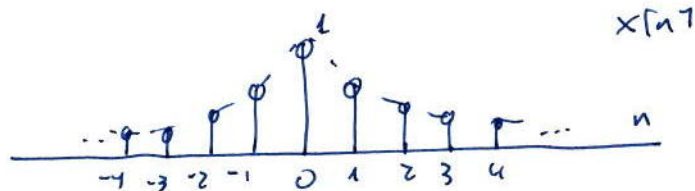
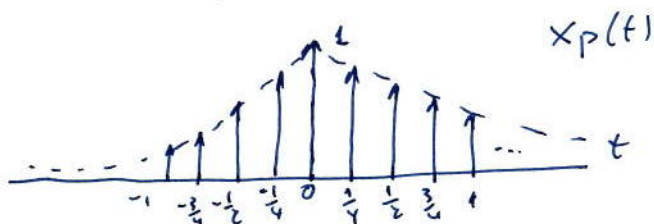
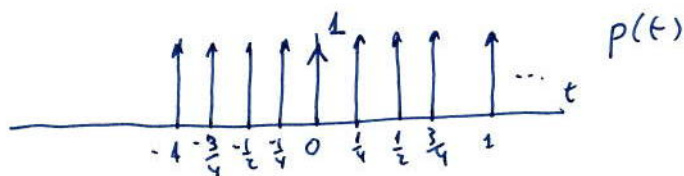
$$\omega_s = 2\omega_{\max} \Rightarrow \omega_s = 10\pi$$

$$T = \frac{1}{5}$$

b)

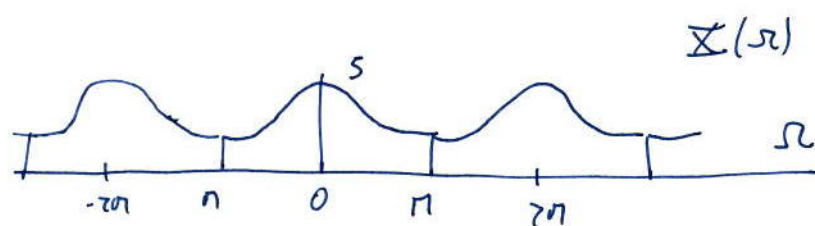
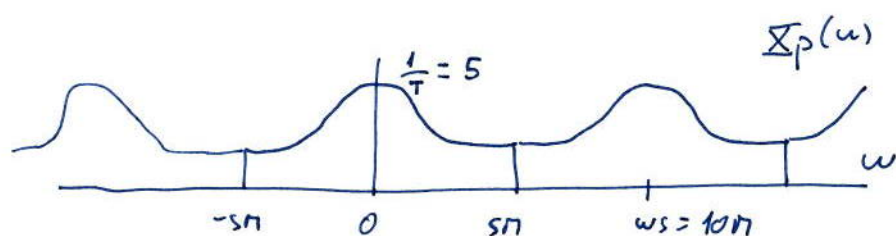
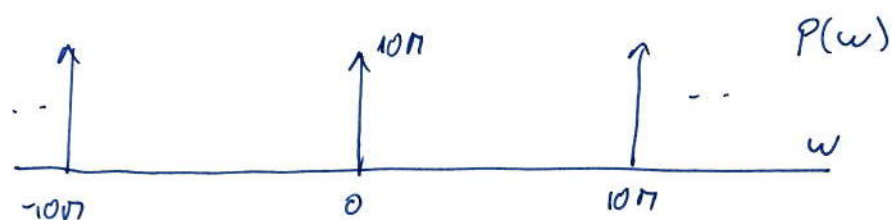
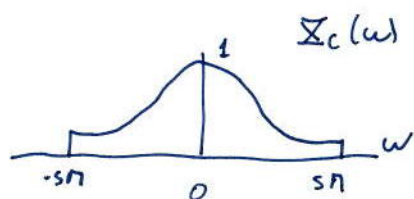


$$T = 1/4$$



$$x[n] = x_c(n \cdot T) = x_c\left(\frac{n}{4}\right) = e^{-2\left|\frac{n}{4}\right|} = e^{-\frac{|n|}{2}}$$

c) $V_{50} \quad T = 1/5$

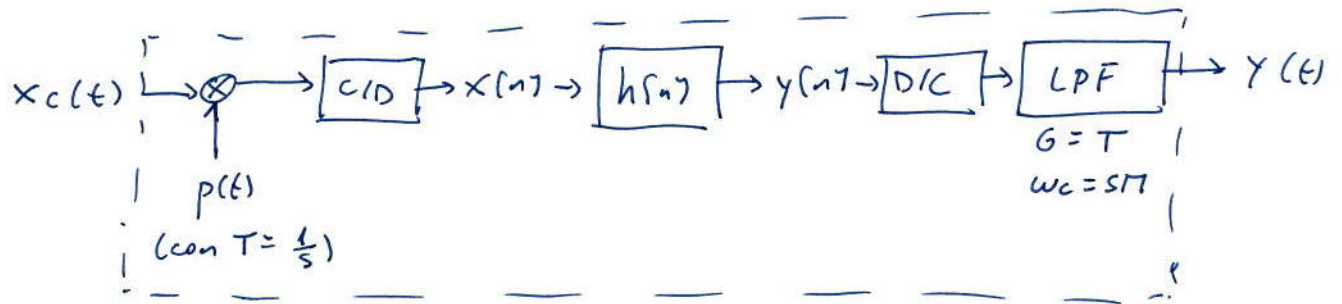


$$X(r) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X_c\left(\frac{r - 2\pi k}{T}\right)$$

For no haber aliasing: $X(r) = \frac{1}{T} X_c\left(\frac{r}{T}\right) \quad |r| < \pi$

$$X(r) = 5 X_c(5r) = \frac{20}{4 + (5r)^2} \quad |r| < \pi$$

d) Esquema:



$$H_c(\omega) = \frac{1}{(1+j\omega)^2}$$

→ ① Límite en Banda

$$H_c(\omega) = \begin{cases} \frac{1}{(1+j\omega)^2} & |\omega| \leq 5\pi \\ 0 & |\omega| > 5\pi \end{cases}$$

② CAMBIO VARIABLE

$$H(\Omega) = \frac{1}{(1+j5\Omega)^2} \quad |\Omega| < \pi$$

↳ (PERIÓDICO DE PERÍODO 2π)